

### XENSIV™ high accuracy coreless current sensor for external current rail applications

### **Features**

- High accurate, scalable, DC & AC current sensing
- Full scale up to ±41.6 mT for current measurements above 1000 A
- 5 V supply voltage
- Ratiometric single-ended analog output
- ISO 26262 Safety Element out of Context for safety requirements up to ASIL B
- Small form factor enables high power density
- Very low sensitivity error over temperature and lifetime
- · Excellent stability of offset over temperature and lifetime
- Differential sensor principle ensures superior magnetic stray field suppression
- · High bandwidth provides fast response time and low phase delay
- Over-Current Detection with configurable threshold (max detection time < 1.0 μs)
- EEPROM with integrated charge-pump for in-system programming and end-of-line calibration without additional circuitry
- · Different package solutions for optimal system integration in different wiring schemes
- Digital control and diagnostic interface. It provides: temperature readout, safety status readout, read/write programming access to internal EEPROM, control of diagnosis mode
- Pre-calibrated over temperature with possibility of in-system end-of-line calibration to compensate system-level errors

### **Potential applications**

- · Main and auxiliary electrical drives
- · Over-load and over-current detection
- · Battery main switch
- Smart circuit breakers
- PV inverters
- Current monitoring and power supplies

### **Product validation**

Qualified for automotive applications. Product validation according to AEC-Q100, Grade 0.

### **Description**

TLE4973 is a high precision miniature coreless magnetic current sensor for AC and DC measurements with analog interface and fast over-current detection output. Negative effects (saturation, hysteresis) known from open loop sensors using flux concentration techniques are avoided. The differential measurement principle allows great stray field suppression for operation in harsh environments. The high configurability enables customization for a variety of applications. User-programmable parameters such as OCD thresholds and sensitivity settings are stored in an embedded EEPROM. Device configuration can be performend in-situ through a one-wire UART based interface called Digital Control Diagnostic Interface (DCDI). The DCDI interface provides read/write access to internal registers of the device and allows to trigger the diagnostic mode and to read temperature, OCD and safety status. It has an auto-addressing functionality and can handle up to 8 slaves on one bus.

Product type	Package	Marking	Ordering code
TLE4973-RE35D5-S0010	PG-TDSO-16	H73E3R0I	SP005448161
TLE4973-RE35S5-S0010	PG-VSON-6	H73E3R0I	SP005448259









## **Datasheet**





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### **Datasheet**

1 Standard configuration



## 1 Standard configuration

## Table 1 TLE4973 standard configuration

Name	RATIO	FS [mT]	FS <sub>5V</sub> [mT]	Sx [mV/mT]	B <sub>THR</sub> [mT]	OCD <sub>GL_MUL</sub>	V <sub>OQ</sub> [V]	<i>OP</i> <sub>MODE</sub>
TLE4973- RE35S5- S0010	an	121 27	126	04.62	120.0	0, filter	W /2	Single-
TLE4973- RE35D5- S0010	on	±21.27	±26	84.63	±29.9	disabled	$V_{\rm DD}/2$	ended AOUT only

RATIO = Ratiometricity to  $V_{\rm DD}$ , FS = minimum Full Scale,  $FS_{\rm SV}$  = Full Scale at  $V_{\rm DD}$  = 5 V, Sx = Sensitivity,  $B_{\rm THR}$  = OCD threshold level,  $OCD_{\rm GL\_MUL}$  = Deglitch filter setting,  $V_{\rm OQ}$  = Quiescent output voltage,  $OP_{\rm MODE}$  = Output mode 1) Configuration can be changed by the customer.

2 Functional block diagram



## 2 Functional block diagram

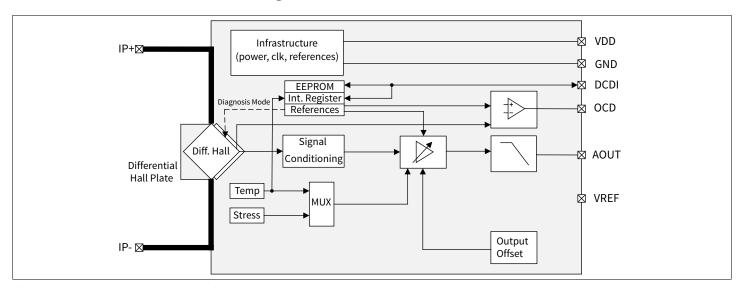


Figure 1 Device functional block diagram

The current flowing through an external current rail induces a magnetic field that is measured by two Hall probes differentially. The differential measurement principle provides superior magnetic stray field suppression. A high-performance amplifier combines the signal resulting from the differential field and the compensation information provided by the temperature and stress compensation unit. Finally, the amplifier output signal is fed into a differential output amplifier which drives the analog output of the sensor.

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3 Pin configuration



# 3 Pin configuration

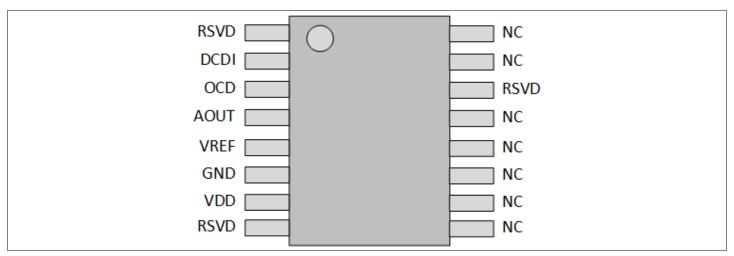


Figure 2 Pin configuration (PG-TDSO-16)

Table 2 Pin definitions and functions (PG-TDSO-16)

Pin No.	Symbol	Function	Comment
1	RSVD	Pin shorted with sensor lead frame	1)
2	DCDI	DCDI communication interface	Open drain input / output, short to VDD if not used
3	OCD	Over Current Detection output	Open drain output, short to GND if not used
4	AOUT	Analog output	-
5	VREF	Not used	Keep open
6	GND	Ground	-
7	VDD	Supply voltage	-
8	RSVD	Pin shorted with sensor lead frame	1)
9-13	NC	Not connected with lead frame or sensor	-
14	RSVD	Pin shorted with sensor lead frame	1)
15-16	NC	Not connected with lead frame or sensor	-

<sup>1)</sup> Lead frame shall be connected to GND. Only one RSVD pin shall be connected to GND to avoid GND loops through the lead frame. The other RSVD pins shall be left open.

### **Datasheet**

3 Pin configuration



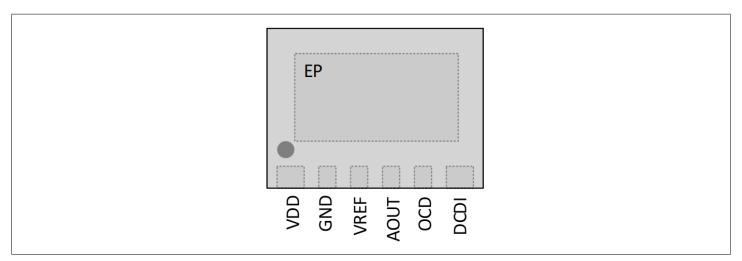


Figure 3 Pin configuration (PG-VSON-6)

Table 3 Pin definitions and functions (PG-VSON-6)

Pin No.	Symbol	Function	Comment
1	VDD	Supply voltage	-
2	GND	Ground	-
3	VREF	Not used	Keep open
4	AOUT	Analog output	-
5	OCD	Over Current Detection output	Open drain output, short to GND if not used
6	DCDI	DCDI communication interface	Open drain input / output, short to VDD if not used
7	EP	Exposed pad	Shall be connected to GND

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4 General product characteristics



## 4 General product characteristics

## 4.1 Absolute maximum ratings

### Table 4 Absolute maximum ratings

Parameter	Symbol		Values		Unit	Note or condition
		Min.	Тур.	Max.		
Supply voltage maximum ratings	V <sub>DD</sub>	-0.3	-	6.5	V	
Voltage on interface pins VREF, AOUT	V <sub>IO_A</sub>	-0.3	-	V <sub>DD</sub>	V	Pull-up resistors are connected to $V_{\rm DD}$
Voltage on interface pins OCD, DCDI	V <sub>IO_D</sub>	-0.3	-	$V_{\mathrm{DD}}$	V	Pull-up resistors of the open-drain drivers are connected to $V_{\rm DD}$
ESD voltage HBM (human body model)	V <sub>ESD_HBM</sub>	-2	-	2	kV	Human Body Model (HBM), according to AEC Q100-002 standard
ESD voltage CDM (charged device model)	V <sub>ESD_CDM</sub>	-1	-	1	kV	Charged Device Model (CDM), according to JEDEC JS-002 standard
Junction temperature maximum ratings	T <sub>J_MAX</sub>	-	-	175	°C	
Storage temperature	T <sub>A_STORE</sub>	-40	_	150	°C	

### Attention:

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the section "functional range" of this data sheet is not implied. Furthermore, only single error cases are assumed. More than one stress/error case may also damage the device. Exposure to absolute maximum rating conditions for extended periods may affect device reliability. During absolute maximum rating overload conditions the voltage on  $V_{\rm DD}$  pins with respect to ground must not exceed the values defined by the absolute maximum ratings. Lifetime statements are an anticipation based on an extrapolation of Infineon's qualification test results. The actual lifetime of a component depends on its form of application and type of use etc. and may deviate from such statement. Lifetime statements shall in no event extend the agreed warranty period.

### **Datasheet**

4 General product characteristics



# 4.2 Functional range

The following operating range must not be exceeded in order to ensure correct operation of the device. All parameters specified in the following sections refer to these operating conditions unless otherwise indicated.

Table 5 Functional range

Parameter	Symbol	Values			Unit	Note or condition
		Min.	Тур.	Max.		
Operating supply voltage	$V_{\mathrm{DD}}$	4.5	5	5.5	V	
Ambient temperature	T <sub>A</sub>	-40	_	150	°C	
Operating junction temperature	$T_{J_{OP}}$	-40	-	160	°C	
Capacitance on AOUT pin	C <sub>AOUT</sub>	6	6.8	8	nF	Without series resistor, including parasitic capacitance on the board
Sensitivity calibration range	S <sub>CAL_MG</sub>	-15	-	15	%	Minimum residual calibration range available for customer sensitivity calibration after Infineon end of production line

### **Datasheet**

5 Product features



## **5** Product features

## **5.1** Electrical characteristics

### Table 6Electrical characteristics

Parameter	Symbol		Values	i	Unit	Note or condition	
		Min.	Тур.	Max.			
Current consumption	I <sub>DD</sub>	-	17.5	21	mA	I <sub>AOUT</sub> = 0 mA	
Programming current	I <sub>PROG</sub>	-	-	25	mA	Total current consumption during EEPROM programming	
Power-on delay time	t <sub>POR</sub>	-	_	1.5	ms	$V_{\rm DD}$ rising 0 V to 5.0 V. Starts when the $V_{\rm DD}$ reaches the minimum allowed operating conditions and stops when the device is fully operational. OCD is set to GND within this time. DCDI pin is set to high ohmic 'Z'	
Quiescent output voltage bidirectional	$V_{\rm OQBID}$	-	V <sub>DD</sub> /2	-	V		
Quiescent output voltage unidirectional	V <sub>OQUNI</sub>	_	V <sub>DD</sub> / 3.984	-	V		
Analog output linear range	V <sub>AOUT_LR</sub>	0.3	_	4.7	V	$V_{\rm DD} = 5 \text{ V}, -6.53 \text{ mA} \le I(AOUT) \le 6.53 \text{ mA}$	
Analog output drive capability	I <sub>AOUT</sub>	-6.53	-	6.53	mA	DC current	
Analog output saturation voltage	V <sub>CL_H</sub>	-	-	300	mV	I(AOUT) = -6.53 mA	
Analog output saturation voltage	V <sub>CL_L</sub>	V <sub>DD</sub> - 0.3	-	-	V	I(AOUT) = 6.53 mA	
Analog output diagnosis mode output amplitude variation	V <sub>DIAG_VAR</sub>	-20	-	20	%	Over temperature and life time with respect to diagnosis mode test signal	

### **Datasheet**

5 Product features



# 5.2 Sensing characteristics

## Table 7 Sensing characteristics

Parameter	Parameter Symbol		Values			Note or condition
		Min.	Тур.	Max.		
Sensitivity, range S1	S1	-	52.89	-	mV/mT	Bidirectional output mode:  • $FS = \pm 34.04 \text{ mT}$ • $FS_{5V} = \pm 41.6 \text{ mT for } V_{DD} = 5 \text{ V}$ Unidirectional output mode:  • $FS = 57.67 \text{ mT}$
Sensitivity, range S2	S2	-	63.47	-	mV/mT	Bidirectional output mode:  • FS = ±28.36 mT  • FS <sub>5V</sub> = ±34.66 mT for V <sub>DD</sub> = 5 V Unidirectional output mode:  • FS = 48.05 mT
Sensitivity, range S3	S3	-	84.63	-	mV/mT	Bidirectional output mode:  • $FS = \pm 21.27 \text{ mT}$ • $FS_{5V} = \pm 26 \text{ mT for } V_{DD} = 5 \text{ V}$ Unidirectional output mode:  • $FS = 36.04 \text{ mT}$
Sensitivity, range S3.5	S3.5	-	105.8	_	mV/mT	Bidirectional output mode:  • FS = ±17.01 mT  • FS <sub>5V</sub> = ±20.79 mT for V <sub>DD</sub> = 5 V  Unidirectional output mode:  • FS = 28.83 mT
Sensitivity, range S4	S4	-	126.95	-	mV/mT	Bidirectional output mode:  • $FS = \pm 14.18 \text{ mT}$ • $FS_{5V} = \pm 17.33 \text{ mT for } V_{DD} = 5 \text{ V}$ Unidirectional output mode:  • $FS = 24.03 \text{ mT}$
Sensitivity, range S4.5	S4.5	-	148.12	-	mV/mT	Bidirectional output mode:  • $FS = \pm 12.15 \text{ mT}$ • $FS_{5V} = \pm 14.85 \text{ mT for } V_{DD} = 5 \text{ V}$ Unidirectional output mode:  • $FS = 20.59 \text{ mT}$
Sensitivity, range S5	S5	-	169.26	_	mV/mT	Bidirectional output mode:  • $FS = \pm 10.64 \text{ mT}$ • $FS_{5V} = \pm 13 \text{ mT for } V_{DD} = 5 \text{ V}$ Unidirectional output mode:  • $FS = 18.02 \text{ mT}$

(table continues...)

### **Datasheet**

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#### (continued) Sensing characteristics Table 7

Parameter Sy	Symbol		Values		Unit	Note or condition
		Min.	Тур.	Max.		
Sensitivity, range S5.5	S5.5	-	190.44	-	mV/mT	Bidirectional output mode:  • FS = ±9.45 mT  • FS <sub>5V</sub> = ±11.55 mT for V <sub>DD</sub> = 5 V  Unidirectional output mode:  • FS = 16.02 mT
Sensitivity, range S6	S6	_	253.89	_	mV/mT	Bidirectional output mode:  • FS = ±7.09 mT  • FS <sub>5V</sub> = ±8.67 mT for V <sub>DD</sub> = 5 V  Unidirectional output mode:  • FS = 12.01 mT
Quiescent ratiometricity factor	K <sub>OQ</sub>	-	1	-	-	
Sensitivity ratiometricity factor	K <sub>S</sub>	-	1	-	-	
Transfer function cutoff frequency	BW	120	210	-	kHz	-3dB criterion, C <sub>AOUT</sub> = 6.8 nF
Output phase delay	arphiDELAY	_	43.2	60.5	0	$f_{SIGNAL} = 120 \; kHz$
Output group delay	$arphi_{G_DELAY}$	_	1.0	1.4	μs	
Output group delay variation	φ G_DELAY_VAR	-20	-	20	%	Over temperature and lifetime
Output noise density	B <sub>NOISE</sub>	-	70	175	nT/√Hz	Referenced to input differential magnetic field. Typical value is for $T_A = 25$ °C. Max value is for $T_A \le 150$ °C <sup>1)</sup>
External homogeneous magnetic field suppression	B <sub>SR</sub>	34	40	-	dB	With respect to the external homogeneous field applied, up to 20 mT. Frequency up to 4 kHz
Initial sensitivity error	E <sub>SENS</sub>	-5.0	±2.7	5.0	%	Typical value is $\pm 3$ sigma. $T_A = 25$ °C; can be compensated in the application by end-of-line calibration at RT.
Sensitivity calibration resolution	E <sub>SENS_CAL</sub>	-	0.12	-	%	Calibration resolution varies with calibration code. Further details to be found in the user manual

(table continues...)

### **Datasheet**

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## Table 7 (continued) Sensing characteristics

Parameter Sym	Symbol		Values		Unit	Note or condition	
		Min.	Тур.	Max.			
Sensitivity drift over temperature	E <sub>SENS_T</sub>	-1.35	±0.8	1.35	%	Bidirectional output mode. Typical is ±3 sigma At 0h. <sup>2)</sup>	
Sensitivity drift over temperature and lifetime	E <sub>SENS_L</sub>	-1.9	±1.2	1.9	%	Bidirectional output mode. $T_A = 25$ °C to $150$ °C. $\frac{2)(3)}{}$	
Initial offset error	E <sub>OFF_INIT</sub>	-200	±80	200	μТ	Typical value is $\pm$ 3 sigma. $T_A$ = 25 °C; can be compensated in the application by end-of-line calibration at RT	
Offset calibration resolution	E <sub>OFF_CAL</sub>	_	12.9	-	μТ	Further details to be found in the user manual	
Output offset variation over temperature	E <sub>OFF_T</sub>	-100	±50	100	μТ	Bidirectional output mode. Typical is ±3 sigma At 0h. <sup>2)</sup>	
Output offset variation over temperature and lifetime	E <sub>OFF_L</sub>	-100	±50	100	μТ	Bidirectional output mode. <sup>2) 3)</sup>	
Total drift over temperature	E <sub>TOT_T</sub>	-1.6	±1	1.6	%	Bidirectional output mode. Typical is ±3 sigma At 0h. <sup>2)</sup> Percentage of full scale, sensitivity <i>S1</i> . Includes sensitivity, offset and linearity errors.	
Total drift over temperature	E <sub>TOT_L</sub>	-2	±1.35	2	%	Bidirectional output mode. $T_A = 25$ °C to $150$ °C. $^{2/3}$	
and lifetime						Percentage of full scale, sensitivity S1. Includes sensitivity, offset and linearity errors.	
Sensitivity ratiometricity error over temperature and lifetime	E <sub>SRATIO_L</sub>	-	±0.05	-	%/%V <sub>DD</sub>	Bidirectional output mode. ±10% V <sub>DD</sub> variation. <sup>3)</sup>	
Offset ratiometricity error over temperature and lifetime	E <sub>ORATIO_L</sub>	-	±6	-	μТ	Bidirectional output mode. $\pm 10\% V_{\rm DD}$ variation. <sup>3)</sup>	

(table continues...)

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#### (continued) Sensing characteristics Table 7

Parameter 5	Symbol	Values			Unit	Note or condition
		Min.	Тур.	Max.		
Temperature sensor sensitivity	S <sub>TEMP</sub>	-	16	-	LSB/°C	
Temperature sensor RMS noise	NRMS <sub>TEMP</sub>	-	-	0.5	°C	In the range of 25 °C to max operating $T_{\rm J}$
Temperature sensor linearity error	E <sub>TEMP</sub>	-5	-	5	°C	In the range of 25 °C to max operating $T_J$

Output noise density can be expressed according to the formula shown in Chapter 5.3.3.

Specified accuracy errors are only valid for the pre-programmed ratiometricity setting.

<sup>1)</sup> 2) 3) Min/Max values derived from AEC-Q100 qualification. Typical values are ±3 sigma values measured during AEC-Q100 qualification.



### 5.2.1 Full scale definition

The full scale input magnetic field range and the corresponding full scale output voltage range on AOUT pin are defined as follows.

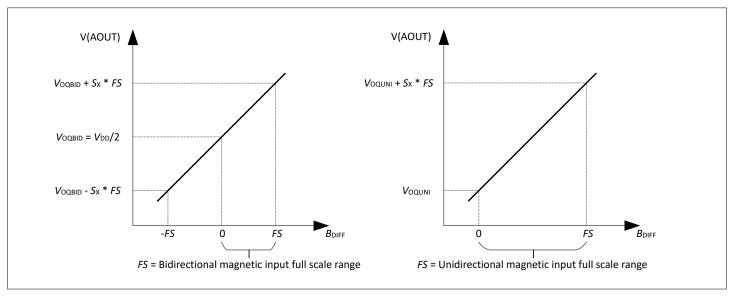


Figure 4 Full scale range definition (FS)

In Table 7 the full scale range FS is defined with respect to the sensitivity  $S_X$ . The total error drift specification in Table 7 considers the full scale range FS to be related to the maximum absolute value measurable by the sensor and not the whole excursion from -FS to +FS. Referring the total error to the whole excursion would reduce the total error, which is defined as % of FS, by factor 2.



# **5.2.2** Typical performance over temperature

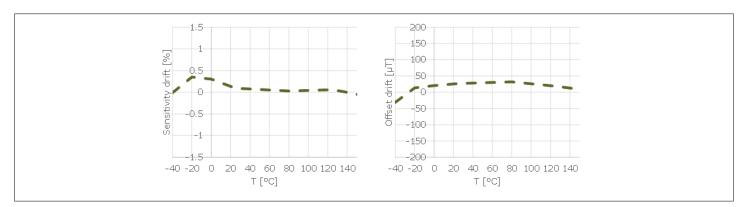


Figure 5 PG-VSON-6, single-ended AOUT only bidirectional, sensitivity S1

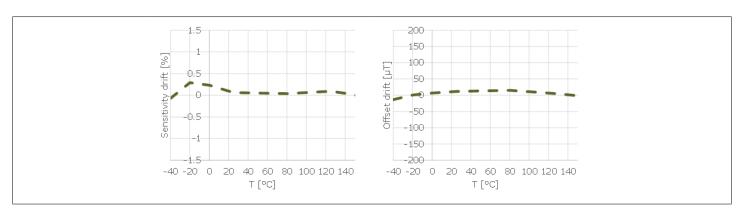


Figure 6 PG-VSON-6, single-ended AOUT only bidirectional, sensitivity S6

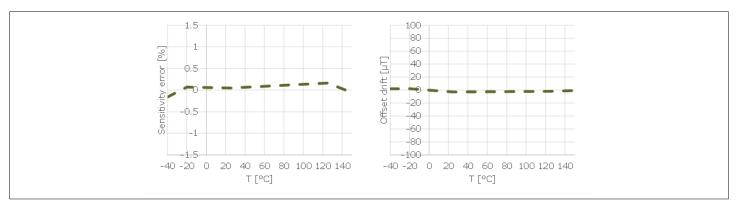


Figure 7 PG-TDSO-16, single-ended AOUT only bidirectional, sensitivity S1

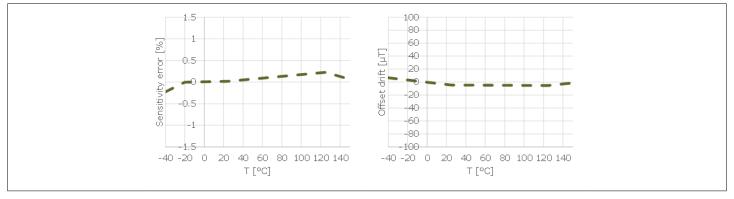


Figure 8 PG-TDSO-16, single-ended AOUT only bidirectional, sensitivity S6



#### 5.3 **Functional Description**

#### 5.3.1 **Output voltage and current polarity**

TLE4973 coreless magnetic current sensor measures the differential magnetic field caused by a current flowing in an external conductor. The figures below show that the magnetic fields  $B_{\rm H1}$  and  $B_{\rm H2}$  at the sensor's Hall cell position depend on the current  $I_{PN}$ . The sensor provides an output voltage with the same polarity of the current  $I_{PN}$ 

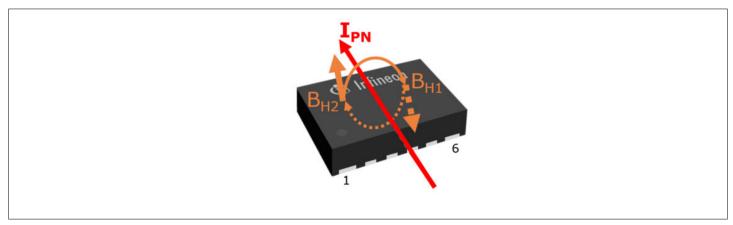


Figure 9 **Current polarity (PG-VSON-6)** 

The current  $I_{PN}$  is positive when it flows in the direction as shown in the figure above through an external current rail.

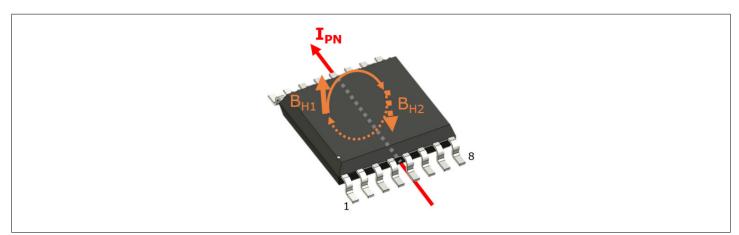


Figure 10 **Current polarity (PG-TDSO-16)** 

The current  $I_{PN}$  is positive when it flows in the direction from pin 1 (+) to pin 9 (-) through an external current rail.

#### 5.3.2 **Output behavior**

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#### 5.3.2.1 Single-ended AOUT only output mode

The AOUT pin operates as an output and provides a single-ended output voltage.

Being S the sensitivity and  $V_{00}$  the quiescent voltage of the sensor, the output voltage on AOUT is described as follows:

$$V(AOUT) = V_{OQ} + S \cdot B_{DIFF} \tag{1}$$

Where  $B_{DIFF}$  is defined as the average differential magnetic field at the Hall probes location:



$$B_{DIFF} = \frac{B_{H1} - B_{H2}}{2} \tag{2}$$

The nominal quiescent voltage value at the pin AOUT is programmable at 2 different values ( $V_{\text{OQBID}}$ ,  $V_{\text{OQUNI}}$ ) and is ratiometric with respect to  $V_{\text{DD}}$  as defined in the following formula:

$$V_{OQ}\left(V_{DD}\right) = V_{OQ}\left(5V\right) \cdot \left[1 + \frac{\left(V_{DD} - 5V\right) \cdot K_{OQ}}{5V}\right] \tag{3}$$

Since nominal value of the ratiometric factor  $K_{OO}$  is 1, the formula becomes:

$$V_{OQ}(V_{DD}) = \frac{V_{DD}}{2} \tag{4}$$

The sensitivity S is ratiometric to  $V_{DD}$  and is expressed as follows:

$$S(V_{DD}) = S(5V) \cdot \left[ 1 + \frac{(V_{DD} - 5V) \cdot K_S}{5V} \right] \tag{5}$$

Since the nominal value of the ratiometric factor is 1 the formula simplifies as follows:

$$S(V_{DD}) = S(5V) \cdot \frac{V_{DD}}{5V} \tag{6}$$

## 5.3.3 Output noise density

Output noise density referenced to input differential magnetic field can be expressed according to the following formula:

$$B_{NOISE} = \frac{V_{NOISE}}{\sqrt{\frac{\pi}{2} \cdot BW}} \cdot \frac{1}{S} \cdot 10^9 \tag{7}$$

Where:

- $B_{\text{NOISE}}$  is the output noise density in [nT/ $\sqrt{\text{Hz}}$ ];
- *V*<sub>NOISE</sub> is the output noise voltage in [VRMS];
- BW is the transfer function cutoff frequency in [Hz];
- S is the sensitivity in [mV/mT].

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### 5.3.4 Safety and diagnostic

### 5.3.4.1 OCD - Over Current Detection

The device provides fast Over Current Detection (OCD) output on OCD pin in case of an over-current event. Over-current condition is present when the input current absolute value (positive or negative) is higher than the chosen threshold level.

## 5.3.4.1.1 OCD pins external connection

The OCD pin operates as active low, open-drain output. This allows easy setup of a wired-AND configuration enabling the monitoring of several current sensors outputs via only one microcontroller pin. The OCD pins can be connected to a logic input pin of the microcontroller and/or the gate-driver, to quickly react to over current events.

The device has an EEPROM option to configure over current detection output to latched or not-latched mode.

In case latched mode is selected, OCD output is permanently latched to low state on over-current event. Latch can be reset via DCDI interface.

In case not-latched mode is selected,OCD output is set to low state on overcurrent event on the current rail and is released when overcurrent event disappears.

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### 5.3.4.1.2 OCD thresholds

The threshold level of the OCD output is programmable. Threshold level is used symmetrically for positive and negative over current events. The possible thresholds levels can be calculated with the formulas below. For sensitivity values between *S1* and *S3.5* the following formula applies:

$$OCD_{THR} = \frac{B_{THR} \cdot 1000}{TF} = \frac{[(OCD_{CODE} \cdot 1.745) + 10.714]_{[mT]} \cdot 1000}{TF}$$
(8)

For sensitivity values between S4 and S6 the following formula applies:

$$OCD_{THR} = \frac{B_{THR} \cdot 1000}{TF} = \frac{[(OCD_{CODE} \cdot 0.698) + 4.286]_{[mT]} \cdot 1000}{TF}$$
(9)

### Where:

- OCD<sub>THR</sub> is the OCD threshold level in [A];
- B<sub>THR</sub> is the nominal OCD threshold level in [mT]
- OCD<sub>CODE</sub> is the decimal value of a 6 bits code. Minimum allowed value for S4-S6 is 5;
- TF is the current rail transfer factor in  $[\mu T/A]$ .

### Table 8 TLE4973 OCD thresholds between S1 and S3.5

OCD <sub>CODE</sub>	B <sub>THR</sub> [mT]	Notes				
0	10.7	Minimum setting				
11	29.9	Pre-programmed setting				
63	121	Maximum setting				

### Table 9 TLE4973 OCD thresholds between S4 and S6

OCD <sub>CODE</sub>	B <sub>THR</sub> [mT]	Notes			
5	7.78	Minimum setting			
63	48.3	Maximum setting			



## 5.3.4.1.3 OCD output timing behavior

OCD output features a deglitch filter with programmable deglitching delay time in order to avoid false OCD triggers due to noise spikes on the current rail. The specified deglitching timing is valid when the input current step overtakes the threshold of at least 10%. As a consequence, the OCD output will only be activated when the overcurrent condition is present for a time interval longer than the chosen deglitching delay. See more detailed explanation in the OCD timing diagram. Available options for deglitching delay include 0 and are listed in the OCD parameters table.

Deglitching delay associated with OCD output is reset as soon as the over-current condition is removed.

Figure below shows the OCD output pin nominal behavior during an overcurrent event and defines important timing quantities for which the limits are listed in the parameters table.

Overcurrent pulse 1 is an overcurrent event, where the duration of the overcurrent condition exceeds the overcurrent response time  $t_{\rm D_OCD}$  + response time jitter  $\Delta t_{\rm D_OCD}$  + deglitch filter time  $t_{\rm DEGLITCH}$ . Overcurrent conditions 2 and 3 are not long enough to trigger the OCD output. Pulse 2 is shorter than the overcurrent response time  $t_{\rm D_OCD}$  and therefore no OCD event is generated. In overcurrent condition 3, the overcurrent condition is longer than the response time  $t_{\rm D_OCD}$  + response time jitter  $\Delta t_{\rm D_OCD}$ , but is not exceeding the glitch filter time  $t_{\rm DEGLITCH}$  and no OCD event is generated.

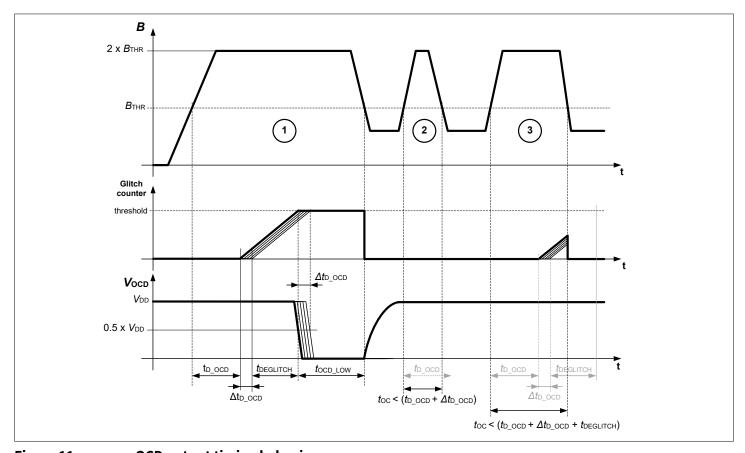


Figure 11 OCD output timing behavior

**Note**: General conditions (unless otherwise specified):  $C_{OCD} = 1 \text{ nF}$ ;  $R_{OCD-PU} = 4.7 \text{ k}\Omega$ 

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### Table 10 OCD characteristics

Parameter	Symbol	Values			Unit	Note or condition
		Min.	Тур.	Max.		
Threshold level tolerance	OCD <sub>THR_TOL</sub>	-10	-	+10	%	
OCD detection minimum time	t <sub>OCD_LOW</sub>	3	-	-	μs	Once the over-current is signaled on the OCD, the signal is kept active for at least this time
OCD response time	$t_{D\_OCD}$	-	0.7	1.0	μs	$B_{\text{DIFF}} = 2 \times B_{\text{THR}}^{1}$
OCD fall time	$t_{F\_OCD}$	_	100	150	ns	From 90% to 10% $V_{\rm DD}$ . Load capacitance $C_{\rm L}$ = 1 nF
OCD response time jitter	$\Delta t_{D\_OCD}$	-	-	0.11	μs	standard deviation, $B_{DIFF} = 2 \times B_{THR}$ , input rise time 1 µs
OCD de-glitch filter basic time	t <sub>OCDGL</sub>	400	500	600	ns	
OCD de-glitch filter setting	OCD <sub>GL_MUL</sub>	0	-	7	_	$t_{\text{DEGLITCH}} = OCD_{\text{GL\_MUL}} \times t_{\text{OCDGL}}; \text{ pre-configured}$ setting = 0 <sup>2</sup>
OCD load capacitance	C <sub>OCD</sub>	_	1	1.3	nF	
OCD pull-up resistor	R <sub>OCD_PU</sub>	0.9	4.7	10	kΩ	To V <sub>DD</sub>
OCD low-side resistance	R <sub>OCD_LOW</sub>	_	_	100	Ω	From OCD pin to GND, when OCD pin is driven low

<sup>1)</sup> Falling edge level of OCD-pin < 0.5 x  $V_{\rm DD}$ .

The device triggers an OCD diagnosis mode when a correspondent command is received through the DCDI interface, in regular operating mode (after the startup time).

It is possible to set the OCD output pin into over current detection mode only. In this mode the OCD pin would not indicate an error except under or overvoltage on the OCD pin.

## 5.3.4.2 Undervoltage / Overvoltage detection

The device activates OCD output and suspends DCDI communication if its own voltage power supply  $(V_{DD})$  is in an under-voltage condition.

Under-voltage condition is present when  $V_{\rm DD}$  is between the values indicated in the  $V_{\rm DD}$  parameter.

In case of overvoltage condition, the sensor provides supply fault indication by disabling the DCDI interface.

Over-voltage condition is present if  $V_{DD}$  higher than the over-voltage threshold parameter ( $V_{OV\ RES}$ ).

<sup>2)</sup> The specified deglitching timing is valid when input current step overtakes the threshold of at least 10%.

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### Table 11 Undervoltage / Overvoltage characteristics

Parameter	Symbol	Values			Unit	Note or condition
		Min.	Тур.	Max.		
Undervoltage/ Overvoltage detection delay	t <sub>UVOV_DELAY</sub>	1	2.4	3	μs	
Undervoltage reset threshold	V <sub>UV_RES</sub>	4.2	_	-	V	V <sub>DD</sub> falling
Undervoltage release threshold	V <sub>UV_REL</sub>	-	-	4.5	V	V <sub>DD</sub> rising
OCD undervoltage condition	$V_{\rm DD\_OCD}$	2.7	-	V <sub>UV_REL</sub>	V	OCD pulled to "low" level
Overvoltage reset threshold	V <sub>OV_RES</sub>	-	-	5.8	V	V <sub>DD</sub> rising
Overvoltage release threshold	V <sub>OV_REL</sub>	5.5	-	-	V	V <sub>DD</sub> falling

## 5.3.4.3 DCDI - Digital Control Diagnostic Interface

The device provides a bus compatible fast one-wire DCDI communication interface on pin DCDI operating as input/output.

The DCDI interface is a single-master, multi-slave UART based interface.

The DCDI pin operates as open-drain.

The device provides a mechanism to detect an internal saturation of the analog output. The detection information is provided on the DCDI interface.

## 5.3.4.3.1 DCDI slave addressing

The device provides an auto-addressing functionality in order to allow the master in the system to address different sensors connected on a bus.

After the device receives the auto addressing broadcast the slave provides its dynamic address (modulated in a voltage level) to the master via the AOUT pin. The master performs the auto addressing broadcast command as long as all sensors provide a unique dynamic address on the dedicated AOUT pin. After the master addresses the dynamic slave address and reassign a static address to each slave, the static address will be stored in the device EEPROM. Up to 8 independent slaves can be addressed by the system.

It is possible that the sensor uses the static address stored in EEPROM, if the EEPROM is configured to use stored address (static). In this case broadcast command to automatically change the address is ignored.

### 5.3.4.3.2 DCDI features

The EEPROM of the device is programmable via the DCDI interface.

It is possible to read the calibration related register content via the DCDI interface.

The DCDI interface protocol layer provides internal errors status information to the master node within each communication frame.

If OCD diagnosis mode or analog output diagnosis mode are activated, the diagnosis mode status bit in the DCDI interface frame will be set to '1'.

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The sensor provides the status of the single internal safety mechanisms if requested by master.

Digital value proportional to the internally measured temperature can be read through the DCDI interface.

The DCDI end-to-end frame protection provides a failure detection rate of more than 99%.

For additional information please refer to the user manual of the device.

Table 12 DCDI characteristics

Parameter	Symbol	Values			Unit	Note or condition
		Min.	Тур.	Max.		
DCDI number of slaves	DCDI <sub>SLV</sub>	-	_	8	-	The number of slaves to be addressed and served in one system
DCDI communication speed	DCDI <sub>SP</sub>	2400	-	57600	bit/s	The baud rate is programmable in 6 different speeds (57600 / 38400 / 19200 / 9600 / 4800 / 2400 bit/s) configurable through 3 bits
DCDI bus voltage	<i>DCDI</i> <sub>VBUS</sub>	-	$V_{\mathrm{DD}}$	-	V	The voltage level for the pull up resistor has the same specification as the sensor supply voltage $V_{\rm DD}$
DCDI low voltage level	V <sub>DCDI_LOW</sub>	_	_	1	V	Low state, transition from high to low
DCDI high voltage level	V <sub>DCDI_HiGH</sub>	1.6	_	_	V	High state, transition from low to high
DCDI pull-up resistor	R <sub>DCDI_PU</sub>	433	-	1320	Ω	
DCDI low-side resistance	R <sub>DCDI_LOW</sub>	-	_	100	Ω	From DCDI pin to GND, when DCDI pin is driven low
DCDI maximum bus load	C <sub>DCDI</sub>	-	-	1.25	nF	Baud rate = 57600 bit/s. For slower baud rates higher capacitance values are allowed.

## 5.3.4.4 Diagnostic mode

The device triggers an analog diagnosis mode when the corresponding command is received through DCDI interface in regular operating mode (after the startup time).

The test signal is programmable at four nominal values.

Table 13 Diagnostic mode

Parameter	Symbol	Values			Unit	Note or condition
		Min.	Тур.	Max.		
Diagnosis mode test amplitude 1		_	5	-	mT	
Diagnosis mode test amplitude 2		_	10	-	mT	
Diagnosis mode test amplitude 3		_	15	-	mT	
Diagnosis mode test amplitude 4		_	20	-	mT	

6 Application Information



## 6 Application Information

## 6.1 Application circuit example for single-ended AOUT only output mode

The figure below shows an example application circuit of the device. Numbers in the application diagram are examples for typical values that could be used in the application.

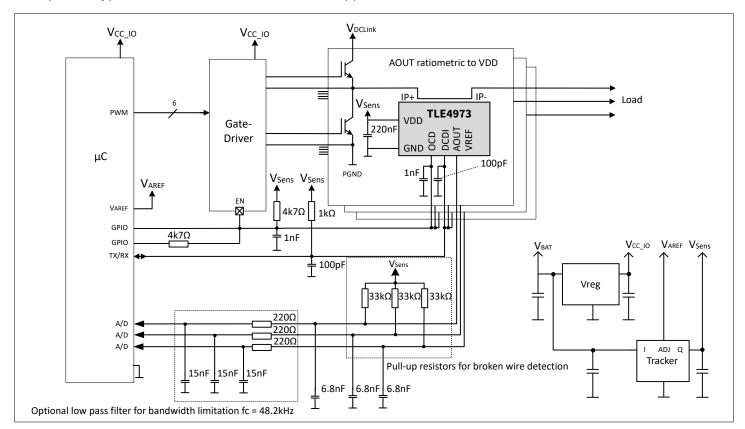


Figure 12 Example application diagram for single-ended AOUT only output mode

Note: This is a simplified example of an application circuit. The function must be verified in the real application.

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7 Package



## 7 Package

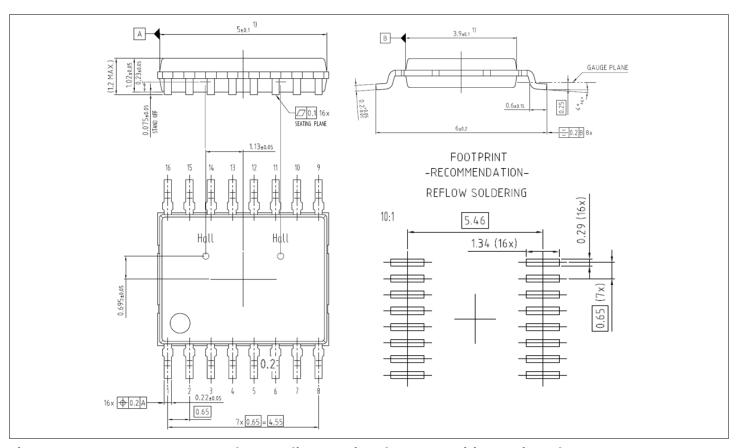


Figure 13 PG-TDSO-16 package outline. Sensing elements position not in scale

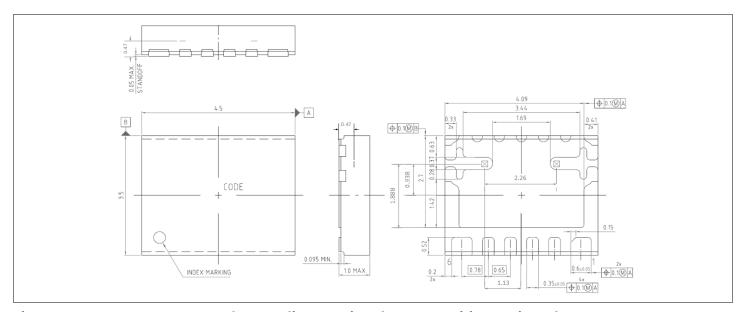


Figure 14 PG-VSON-6 package outline. Sensing elements position not in scale

Table 14 Package grade and moisture sensitivity level according to AEC Q100 standard

Package	Grade	MSL
PG-VSON-6	0	3
PG-TDSO-16	0	3

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7 Package



The package marking of the device in PG-TDSO-16 package is as shown in the figures below.

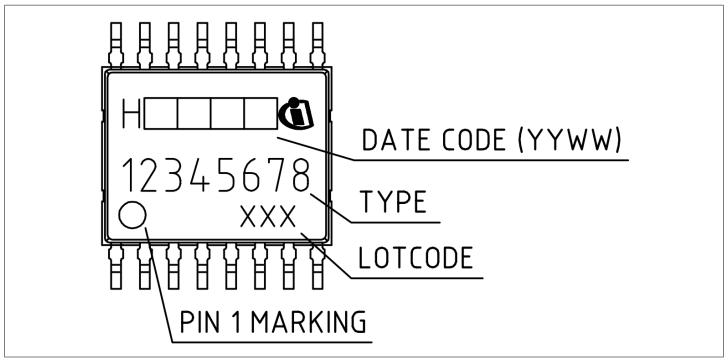


Figure 15 Package marking front side (PG-TDSO-16)

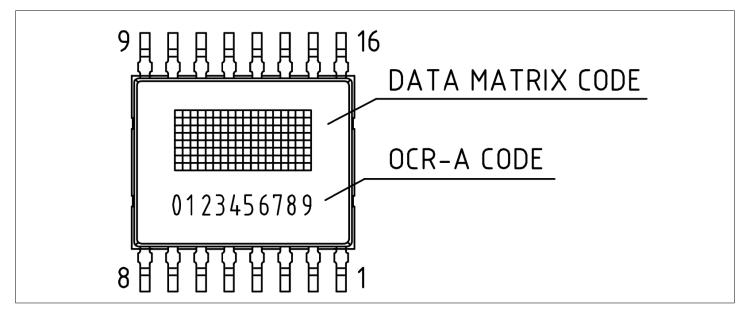


Figure 16 Package marking back side (PG-TDSO-16)

The package marking of the device in PG-VSON-6 is as shown in the figure below.

7 Package



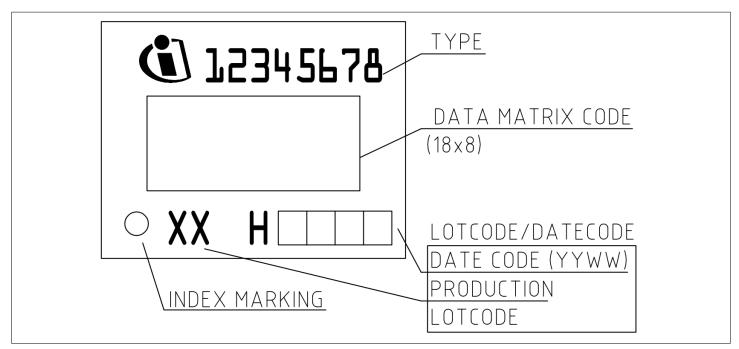


Figure 17 Package marking (PG-VSON-6)

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8 Revision History



# 8 Revision History

## Table 15 Revision History

<b>Revision number</b>	Date of release	Description of changes			
2.2 2025-06-02		<ul><li>Introduced analog output linear range definition;</li><li>Editorial changes.</li></ul>			
2.1	2023-11-30	Updated application diagram examples			
2.0	2023-07-04	<ul> <li>Updated pin configuration: the correct pin configuration in case DCDI is not utilized is to connect the DCDI pin to VDD;</li> <li>Editorial changes.</li> </ul>			
1.0	2023-03-13	Initial release			

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